

## Lecture 7: Audio compression and coding

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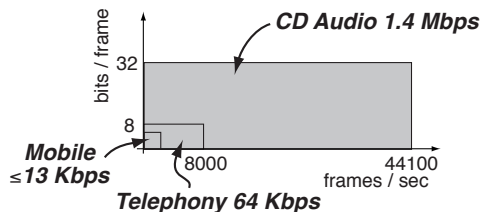
- 1 Information, Compression & Quantization
- 2 Speech coding
- 3 Wide-Bandwidth Audio Coding

# Outline

- 1 Information, Compression & Quantization
- 2 Speech coding
- 3 Wide-Bandwidth Audio Coding

# Compression & Quantization

- How **big** is audio data? What is the **bitrate**?
  - ▶  $F_s$  frames/second (e.g. 8000 or 44100)
  - ×  $C$  samples/frame (e.g. 1 or 2 channels)
  - ×  $B$  bits/sample (e.g. 8 or 16)
  - $F_s \cdot C \cdot B$  bits/second (e.g. 64 Kbps or 1.4 Mbps)

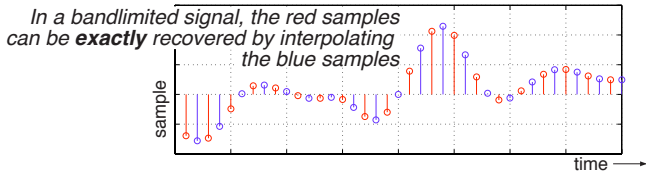


- How to reduce?
  - ▶ lower **sampling rate** → less bandwidth (muffled)
  - ▶ lower **channel count** → no stereo image
  - ▶ lower **sample size** → quantization noise
- Or: use **data compression**

# Data compression:

## Redundancy vs. Irrelevance

- Two main principles in compression:
    - ▶ remove **redundant** information
    - ▶ remove **irrelevant** information
  - **Redundant** information is implicit in remainder
    - ▶ e.g. signal bandlimited to 20kHz, but sample at 80kHz
- can recover every other sample by interpolation:



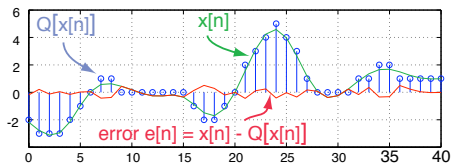
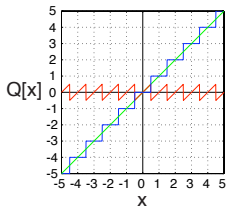
- **Irrelevant** info is unique but unnecessary
  - ▶ e.g. recording a microphone signal at 80 kHz sampling rate

# Irrelevant data in audio coding

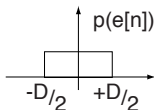
- For coding of audio signals, irrelevant means perceptually insignificant
  - ▶ an empirical property
- Compact Disc standard is adequate:
  - ▶ 44 kHz sampling for 20 kHz bandwidth
  - ▶ 16 bit linear samples for  $\sim 96$  dB peak SNR
- Reflect limits of human sensitivity:
  - ▶ 20 kHz bandwidth, 100 dB intensity
  - ▶ sinusoid phase, detail of noise structure
  - ▶ dynamic properties - hard to characterize
- Problem: separating salient & irrelevant

# Quantization

- Represent waveform with discrete levels



- Equivalent to adding error  $e[n]$ :  
$$x[n] = Q[x[n]] + e[n]$$
- $e[n] \sim$  uncorrelated, uniform white noise



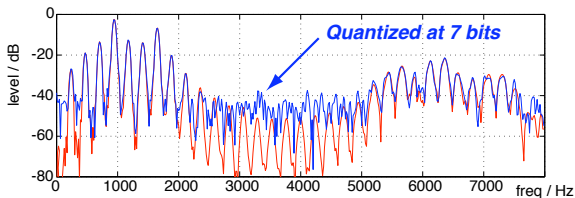
$$\rightarrow \text{variance } \sigma_e^2 = \frac{D^2}{12}$$

# Quantization noise (Q-noise)

- Uncorrelated noise has flat spectrum
  - With a  $B$  bit word and a quantization step  $D$ 
    - ▶ max signal range ( $x$ ) =  $-(2^{B-1}) \cdot D \dots (2^{B-1} - 1) \cdot D$
    - ▶ quantization noise ( $e$ ) =  $-D/2 \dots D/2$
- Best signal-to-noise ratio (power)

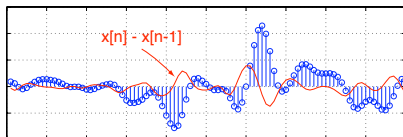
$$\begin{aligned} SNR &= E[x^2]/E[e^2] \\ &= (2^B)^2 \end{aligned}$$

.. or, in dB,  $20 \cdot \log_{10} 2 \cdot B \approx 6 \cdot B$  dB



# Redundant information

- Redundancy removal is *lossless*
- Signal correlation implies redundant information
  - ▶ e.g. if  $x[n] = x[n-1] + v[n]$   
 $x[n]$  has a greater amplitude range  
→ uses more bits than  $v[n]$
  - ▶ sending  $v[n] = x[n] - x[n-1]$  can reduce **amplitude**, hence **bitrate**



- ▶ but: 'white noise' sequence has no redundancy ...
- Problem: separating **unique** & **redundant**



# Optimal coding

- Shannon **information**:

An unlikely occurrence is more 'informative'

$$p(A) = 0.5 \quad p(B) = 0.5$$

**ABBBBAAABBABBABBABB**

**A, B** equiprobable

$$p(A) = 0.9 \quad p(B) = 0.1$$

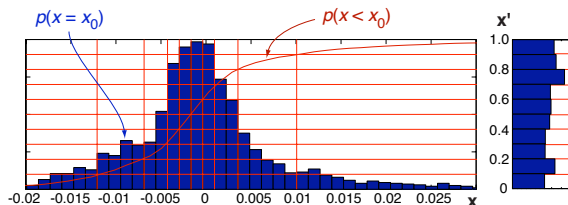
**AAAAABBAAAAAABAAAAAB**

**A** is expected;  
**B** is 'big news'

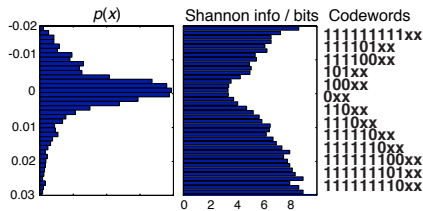
- **Information** in bits  $I = -\log_2(\textit{probability})$ 
  - ▶ clearly works when all possibilities equiprobable
- Optimal bitrate  $\rightarrow$  av.token length = **entropy**  $H = E[I]$ 
  - ▶ .. equal-length tokens are equally likely
- How to achieve this?
  - ▶ transform signal to have uniform pdf
  - ▶ nonuniform quantization for equiprobable tokens
  - ▶ variable-length tokens  $\rightarrow$  Huffman coding

# Quantization for optimum bitrate

- Quantization should reflect pdf of signal:



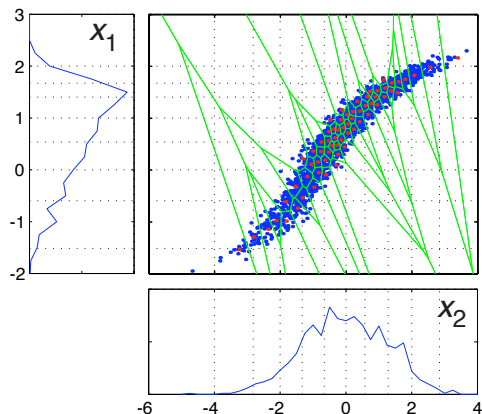
- cumulative pdf  $p(x < x_0)$  maps to uniform  $x'$
- Or, codeword length per Shannon  $\log_2(p(x))$ :



- Huffman coding: tree-structured decoder

# Vector Quantization

- Quantize mutually dependent values in joint space:



- May help even if values are largely independent
  - larger space  $x_1, x_2$  is easier for Huffman

# Compression & Representation

- As always, success depends on **representation**
- Appropriate domain may be 'naturally' bandlimited
  - ▶ e.g. vocal-tract-shape coefficients
  - ▶ can reduce sampling rate without data loss
- In right domain, **irrelevance** is easier to 'get at'
  - ▶ e.g. STFT to separate magnitude and phase

## Aside: Coding standards

- Coding only useful if recipient knows the code!
- **Standardization** efforts are important
- Federal Standards: Low bit-rate secure voice:
  - ▶ FS1015e: LPC-10 2.4 Kbps
  - ▶ FS1016: 4.8 Kbps CELP
- ITU G.x series (also H.x for video)
  - ▶ G.726 ADPCM
  - ▶ G.729 Low delay CELP
- MPEG
  - ▶ MPEG-Audio layers 1,2,3 (mp3)
  - ▶ MPEG 2 Advanced Audio Codec (AAC)
  - ▶ MPEG 4 Synthetic-Natural Hybrid Codec
- More recent 'standards'
  - ▶ proprietary: WMA, Skype...
  - ▶ Speex ...

# Outline

- 1 Information, Compression & Quantization
- 2 **Speech coding**
- 3 Wide-Bandwidth Audio Coding

# Speech coding

- Standard voice channel:
  - ▶ analog: 4 kHz slot ( $\sim 40$  dB SNR)
  - ▶ digital: 64 Kbps = 8 bit  $\mu$ -law  $\times$  8 kHz

- How to compress?

## Redundant

- ▶ signal assumed to be a single voice, not any possible waveform

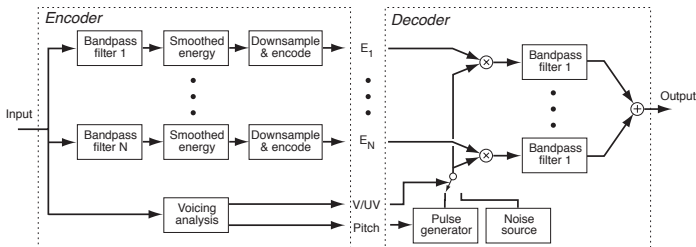
## Irrelevant

- ▶ need code only enough for intelligibility, speaker identification (c/w analog channel)

- Specifically, **source-filter** decomposition
  - ▶ vocal tract &  $f_0$  change slowly
- Applications:
  - ▶ live communications
  - ▶ offline storage

# Channel Vocoder (1940s-1960s)

- Basic **source-filter** decomposition
  - ▶ filterbank breaks into spectral bands
  - ▶ transmit **slowly-changing** energy in each band

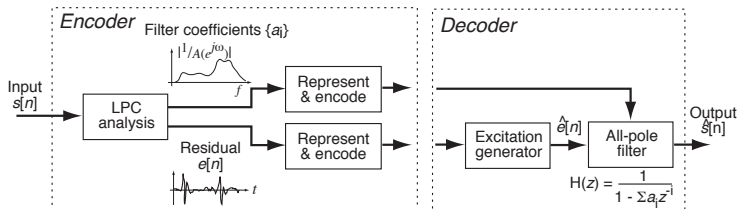


- ▶ 10-20 bands, perceptually spaced
- Downsampling?
- Excitation?
  - ▶ pitch / noise model
  - ▶ or: baseband + 'flattening'...



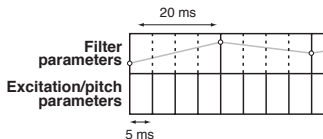
# LPC encoding

- The classic source-filter model



- Compression gains:

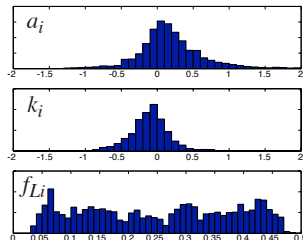
- ▶ **filter parameters** are  $\sim$ slowly changing
- ▶ **excitation** can be represented many ways



# Encoding LPC filter parameters

- For 'communications quality':
  - ▶ 8 kHz sampling (4 kHz bandwidth)
  - ▶  $\sim 10$ th order LPC (up to 5 pole pairs)
  - ▶ update every 20-30 ms  $\rightarrow$  300 - 500 param/s
- Representation & quantization

- ▶  $\{a_i\}$  - poor distribution, can't interpolate
- ▶ reflection coefficients  $\{k_i\}$ : guaranteed stable
- ▶ **LSPs** - lovely!



- Bit allocation (filter):
  - ▶ **GSM** (13 kbps): 8 LARs  $\times$  3-6 bits / 20 ms = 1.8 Kbps
  - ▶ **FS1016** (4.8 kbps): 10 LSPs  $\times$  3-4 bits / 30 ms = 1.1 Kbps

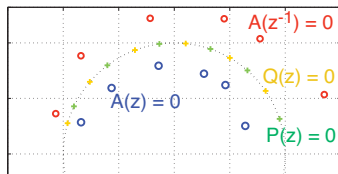
# Line Spectral Pairs (LSPs)

- LSPs encode LPC filter by a set of **frequencies**
- Excellent for quantization & interpolation
- Definition: **zeros** of

$$P(z) = A(z) + z^{-p-1} \cdot A(z^{-1})$$

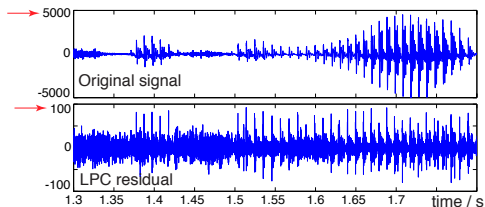
$$Q(z) = A(z) - z^{-p-1} \cdot A(z^{-1})$$

- ▶  $z = e^{j\omega} \rightarrow z^{-1} = e^{-j\omega} \rightarrow |A(z)| = |A(z^{-1})|$  on u.circ.
- ▶  $P(z)$ ,  $Q(z)$  have (interleaved) zeros when  
 $\angle\{A(z)\} = \pm\angle\{z^{-p-1}A(z^{-1})\}$
- ▶ reconstruct  $P(z)$ ,  $Q(z) = \prod_i (1 - \zeta_i z^{-1})$  etc.
- ▶  $A(z) = [P(z) + Q(z)]/2$

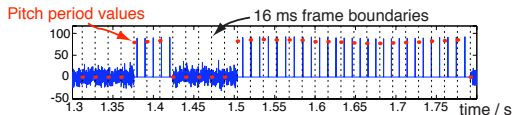


# Encoding LPC excitation

- **Excitation** already better than raw signal:



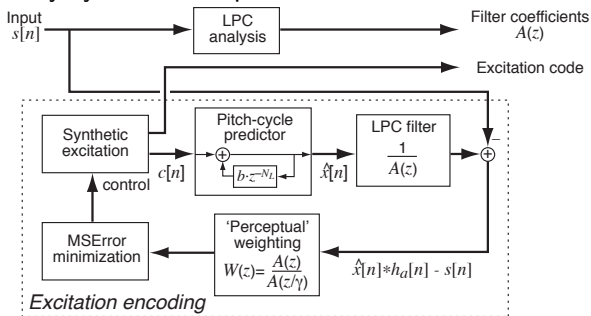
- ▶ save several bits/sample, but still  $> 32$  Kbps
- Crude model: U/V flag + pitch period
  - ▶  $\sim 7$  bits / 5 ms = 1.4 Kbps  $\rightarrow$  LPC10 @ 2.4 Kbps



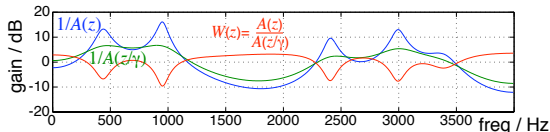
- Band-limit then re-extend (RELPC)

## Encoding excitation

- Something between full-quality residual (32 Kbps) and pitch parameters (1.4 kbps)?
- 'Analysis by synthesis' loop:

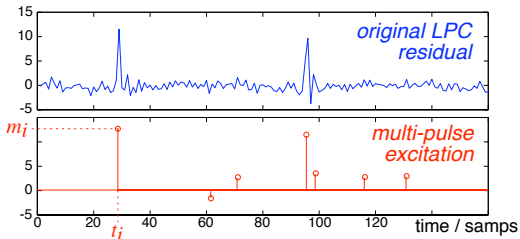


- 'Perceptual' weighting discounts peaks:



# Multi-Pulse Excitation (MPE-LPC)

- Stylize excitation as  $N$  discrete pulses



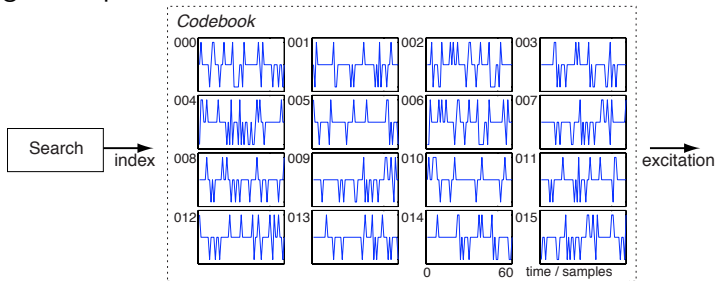
- ▶ encode as  $N \times (t_i, m_i)$  pairs
- Greedy algorithm places one pulse at a time:

$$\begin{aligned} E_{pcp} &= \frac{A(z)}{A(z/\gamma)} \left[ \frac{X(z)}{A(z)} - S(z) \right] \\ &= \frac{X(z)}{A(z/\gamma)} - \frac{R(z)}{A(z/\gamma)} \end{aligned}$$

- ▶  $R(z)$  is residual of target waveform after inverse-filtering
- ▶ cross-correlate  $h_\gamma$  and  $r * h_\gamma$ , iterate

# CELP

- Represent excitation with **codebook**  
e.g. 512 sparse excitation vectors



- ▶ linear search for minimum weighted error?
- **FS1016** 4.8 Kbps CELP (30ms frame = 144 bits):

10 LSPs             $4 \times 4 + 6 \times 3$  bits =    34 bits

Pitch delay         $4 \times 7$  bits =            28 bits

Pitch gain          $4 \times 5$  bits =            20 bits

Codebk index       $4 \times 9$  bits =            36 bits

Codebk gain        $4 \times 5$  bits =            20 bits

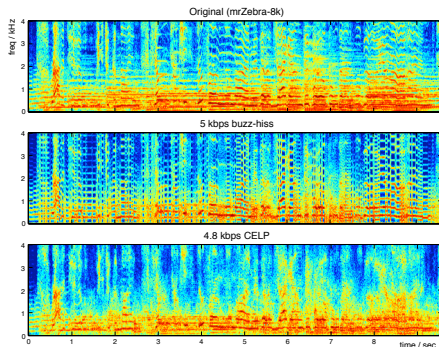
▶ 138 bits

## Aside: CELP for nonspeech?

- CELP is sometimes called a 'hybrid' coder:
  - ▶ originally based on source-filter voice model
  - ▶ CELP residual is waveform coding (no model)

- CELP does not break with multiple voices etc.

- ▶ just does the best it can



- LPC filter models vocal tract; also matches auditory system?
  - ▶ i.e. the 'source-filter' separation is good for relevance as well as redundancy?

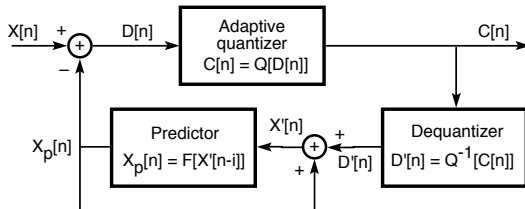


# Outline

- 1 Information, Compression & Quantization
- 2 Speech coding
- 3 Wide-Bandwidth Audio Coding**

# Wide-Bandwidth Audio Coding

- Goals:
  - ▶ **transparent** coding i.e. no perceptible effect
  - ▶ general purpose - handles any signal
- Simple approaches (**redundancy** removal)
  - ▶ Adaptive Differential PCM (ADPCM)

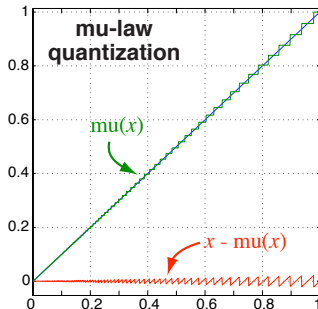


- ▶ as prediction gets smarter, becomes LPC  
e.g. **shorten** - lossless LPC encoding
- Larger compression gains needs **irrelevance**
  - ▶ hide **quantization noise** with psychoacoustic **masking**

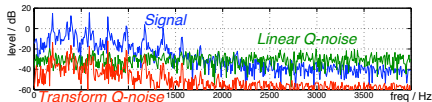
# Noise shaping

- Plain **Q-noise** sounds like added white noise
  - ▶ actually, not all that disturbing
  - ▶ .. but worst-case for exploiting **masking**

- Have Q-noise scale with **signal level**
  - ▶ i.e. quantizer step gets larger with amplitude
  - ▶ minimum distortion for some center-heavy pdf

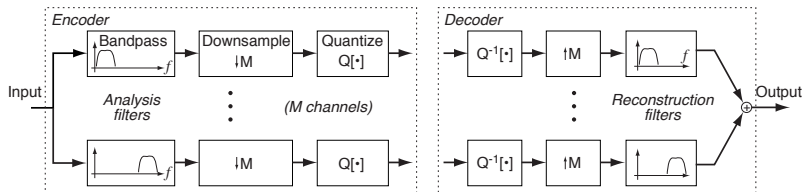


- Or: put Q-noise around peaks in **spectrum**
  - ▶ key to getting benefit of perceptual masking

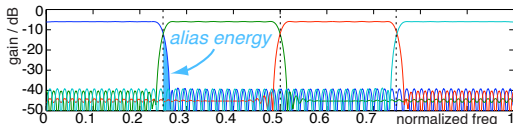


# Subband coding

- Idea: Quantize separately in separate bands



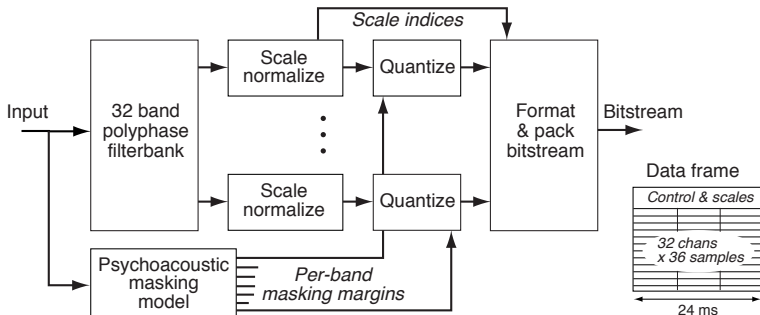
- ▶ **Q-noise** stays within band, gets **masked**
- ‘Critical sampling’  $\rightarrow 1/M$  of spectrum per band



- ▶ some aliasing inevitable
- Trick is to cancel with alias of adjacent band  
 $\rightarrow$  ‘quadrature-mirror’ filters

# MPEG-Audio (layer I, II)

- Basic idea: Subband coding plus psychoacoustic **masking** model to choose **dynamic Q-levels** in subbands



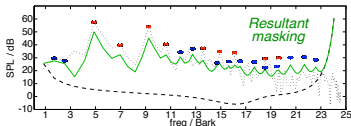
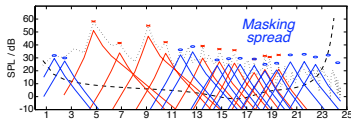
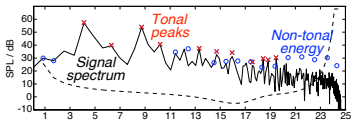
- ▶  $22 \text{ kHz} \div 32 \text{ equal bands} = 690 \text{ Hz bandwidth}$
- ▶  $8 / 24 \text{ ms frames} = 12 / 36 \text{ subband samples}$
- ▶ fixed bitrates 32 - 256 Kbps/ch (1-6 bits/samp)
- ▶ scale factors are like LPC envelope?

# MPEG Psychoacoustic model

- Based on simultaneous masking experiments
- Difficulties:
  - ▶ noise energy masks  $\sim 10$  dB better than tones
  - ▶ masking level nonlinear in frequency & intensity
  - ▶ complex, dynamic sounds not well understood

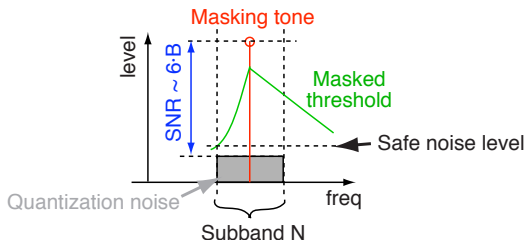
## • Procedure

- ▶ pick 'tonal peaks' in NB FFT spectrum
- ▶ remaining energy  $\rightarrow$  'noisy' peaks
- ▶ apply nonlinear 'spreading function'
- ▶ sum all masking & threshold in power domain



# MPEG Bit allocation

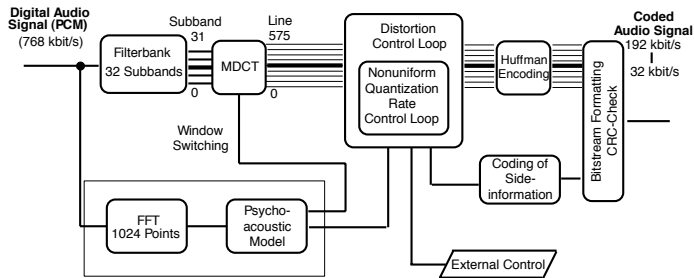
- Result of psychoacoustic model is **maximum tolerable noise** per subband



- ▶ safe noise level → required **SNR** → bits **B**
- Bit allocation procedure (fixed bit rate):
  - ▶ pick channel with worst noise-masker ratio
  - ▶ improve its quantization by one step
  - ▶ repeat while more bits available for this frame
- Bands with **no signal** above masking curve can be skipped

# MPEG Audio Layer III

- 'Transform coder' on top of 'subband coder'

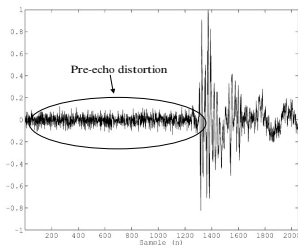


- Blocks of 36 subband time-domain samples become 18 pairs of frequency-domain samples
  - ▶ more **redundancy** in spectral domain
  - ▶ finer control e.g. of aliasing, masking
  - ▶ scale factors now in band-blocks
- Fixed Huffman tables optimized for audio data
- Power-law **quantizer**

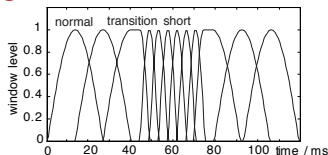


## Adaptive time window

- Time window relies on **temporal masking**
  - ▶ single quantization level over 8-24 ms window
- 'Nightmare' scenario:

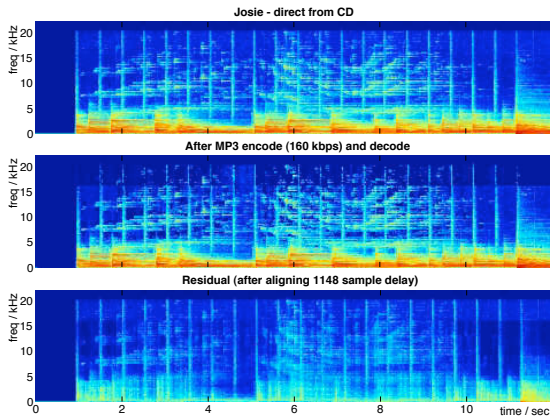


- ▶ 'backward masking' saves in most cases
- **Adaptive switching** of time window:



# The effects of MP3

- Before & after:



- ▶ chop off high frequency (above 16 kHz)
- ▶ occasional other time-frequency 'holes'
- ▶ quantization noise under signal

# MP3 & Beyond

- MP3 is 'transparent' at  $\sim 128$  Kbps for stereo (11x smaller than 1.4 Mbps CD rate)
  - ▶ only **decoder** is standardized:  
better psychological models  $\rightarrow$  better **encoders**
- MPEG2 AAC
  - ▶ rebuild of MP3 without backwards compatibility
  - ▶ 30% better (stereo at 96 Kbps?)
  - ▶ multichannel etc.
- MPEG4-Audio
  - ▶ wide range of component encodings
  - ▶ MPEG Audio, LSPs, ...
- SAOL
  - ▶ 'synthetic' component of MPEG-4 Audio
  - ▶ complete DSP/computer music language!
  - ▶ how to **encode** into it?

# Summary

- For coding, every bit counts
  - ▶ take care over quantization domain & effects
  - ▶ Shannon limits...
- Speech coding
  - ▶ LPC modeling is old but good
  - ▶ CELP residual modeling can go beyond speech
- Wide-band coding
  - ▶ noise shaping 'hides' quantization noise
  - ▶ detailed psychoacoustic models are key

# References

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